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Ceramic Resonator (CERALOCK®)

Application Manual

muRata *Innovator in Electronics*

**Murata
Manufacturing Co., Ltd.**

Cat.No.P17E-19

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Introduction

Ceramic resonators (CERALOCK®) are made of high stability piezoelectric ceramics that function as a mechanical resonator.

This device has been developed to function as a reference signal generator and the frequency is primarily adjusted by the size and thickness of the ceramic element.

With the advance of the IC technology, various equipment may be controlled by a single LSI integrated circuit, such as the one-chip microprocessor. CERALOCK® can be used as the timing element in most microprocessor based equipment.

In the future, more and more applications will use CERALOCK® because of its high stability non-adjustment performance, miniature size and cost savings. Typical applications include TVs, VCRs, automotive electronic devices, telephones, copiers, cameras, voice synthesizers, communication equipment, remote controls and toys.

This manual describes CERALOCK® and will assist you in applying it effectively.

* CERALOCK® is the brand name of these MURATA products.

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1 Characteristics and Types of CERALOCK®

1. General Characteristics of CERALOCK®

Ceramic resonators use the mechanical resonance of piezoelectric ceramics. (Generally, lead zirconium titanate: PZT.)
The oscillation mode varies with resonant frequency. The table on the right shows this relationship.
As a resonator device, quartz crystal is well-known. RC oscillation circuits and LC oscillation circuits are also used to produce electrical resonance. The following are the characteristics of CERALOCK®.

① High stability of oscillation frequency:
Oscillation frequency stability is between that of the quartz crystal and LC or RC oscillation circuits. The temperature coefficient of quartz crystal is $10^{-6}/°C$ maximum and approximately 10^{-3} to $10^{-4}/°C$ for LC or RC oscillation circuits. For comparison these, it is $10^{-5}/°C$ at −20 to +80°C for ceramic resonators.

② Small configuration and light weight:
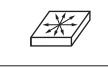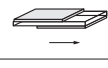The ceramic resonator is half the size of popular quartz crystals.

③ Low price, non-adjustment:
CERALOCK® is mass produced, resulting in low cost and high stability.
Unlike RC or LC circuits, ceramic resonators use mechanical resonance. This means it is not basically affected by external circuits or by the fluctuation of the supply voltage.
Highly stable oscillation circuits can therefore be made without the need of adjustment.
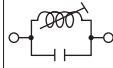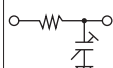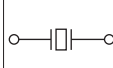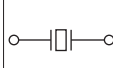
The table briefly describes the characteristics of various oscillator elements.

■ Vibration Mode and Frequency Range

Frequency (Hz) / Vibration Mode		1k	10k	100k	1M	10M	100M	1G
1 Flexural mode		▬	▬					
2 Length mode				▬				
3 Area expansion mode					▬			
4 Radius vibration					▬			
5 Shear thickness mode						▬		
6 Thickness expansion mode						▬	▬	
7 Surface acoustic wave							▬	▬

[Note] : ←——→ show the direction of vibration

■ Characteristics of Various Oscillator Elements

Name	Symbol	Price	Size	Adjust-ment	Oscillation Frequency Initial Tolerance	Long-term Stability
LC		lower cost	Big	Required	±2.0%	Fair
CR		lower cost	Small	Required	±2.0%	Fair
Quartz Crystal		Expen-sive	Big	Not required	±0.001%	Excellent
Ceramic Resonator		Inexpen-sive	Small	Not required	±0.5%	Excellent

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Characteristics and Types of CERALOCK® **1**

2. Types of CERALOCK®

MHz Band lead CERALOCK® (CSTLS Series)

As CSTLS series does not require externally mounted capacitors, the number of components can be reduced, allowing circuits to be made more compact.
The table shows the frequency range and appearance of the three-terminal CERALOCK® with built-in load capacitance.

■ **Part Numbers and Dimensions of lead CERALOCK® (CSTLS Series)**

Part Number	Frequency	Dimensions (in mm)
CSTLS☐G	3.40–10.00MHz	
CSTLS☐X	16.00–70.00MHz	

∗ 16.00–32.99MHz : 3.5

■ **Part Numbering**

(Ex.) | CS | T | LS | 4M00 | G | 5 | 3 | ☐ | -A0 |
      ❶   ❷   ❸   ❹   ❺   ❻   ❼   ❽   ❾

❶ Product ID
❷ Frequency/Built-in Capacitance
❸ Structure/Size
   LS: Round Lead Type
❹ Nominal Center Frequency
❺ Type
   G: Thickness Shear vibration,
   X: Thickness Longitudinal Vibration (3rd overtone)
❻ Frequency Tolerance
   1: ±0.1%, 2: ±0.2%, 3: ±0.3%, 5: ±0.5%, D: DTMF,
   Z: Others
❼ Built-in Load capacitance
   1: 5pF, 3: 15pF, 4: 22pF, 5: 30pF, 6: 47pF
❽ Individual Specification
   With standard products, "❽ Individual Specification" is omitted, and "❾ Package Specification Code" is carried up.
❾ Packaging
   –B0: Bulk,
   –A0: Radial Taping H0=18mm Ammo Pack (Standard)

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1 | Characteristics and Types of CERALOCK®

MHz Band Chip CERALOCK® (CSACW/CSTCC/CSTCR/CSTCE/CSTCW Series)

The MHz band Chip CERALOCK® has a wide frequency range and small footprint to meet further downsizing and high-density mounting requirements.

The table shows the dimensions and two-terminals standard land patterns of the CERALOCK® CSACW series.

The second table shows the dimensions and three-terminals standard land patterns of CSTCC/CSTCR/CSTCE/CSTCW series chip resonator (built-in load capacitance type.) The carrier tape dimensions of CSTCR series are shown on the next page.

■Part Numbering

(Ex.) | CS | T | CR | 4M00 | G | 5 | 3 | | -R0 |
       | ❶ | ❷ | ❸ | ❹ | ❺ | ❻ | ❼ | ❽ | ❾ |

❶ Product ID
❷ Frequency/No capacitance built-in
   A: No Capacitance Built-in, T: Built-in Capacitance
❸ Structure/Size
   CC/CR/CE: Cap Chip Type, CW: Monolithic Chip Type
❹ Nominal Center Frequency
❺ Type
   G: Thickness Shear Vibration,
   V: Thickness Longitudinal Vibration,
   X: Thickness Longitudinal Vibration (3rd overtone)
❻ Frequency Tolerance
   1: ±0.1%, 2: ±0.2%, 3: ±0.3%, 5: ±0.5%, Z: Others
❼ Load Capacitance Value
   (In case of CSACW, value is for external capacitance of standard circuit)
   1: 5pF or 6pF, 2 : 10pF, 3: 15pF, 5: 33pF or 39pF,
   6: 47pF
❽ Individual Specification
   With standard products, " ❽ Individual Specification" is omitted, and " ❾ Package Specification Code" is carried up.
❾ Packaging
   –B0: Bulk,
   –R0: Plastic Taping φ180mm Reel Package

■Dimensions and Standard Land Pattern of Chip CERALOCK® (CSACW Series)

Part Number	Frequency (MHz)	Dimensions Standard Land Pattern (in mm)
CSACW☐X	20.01–70.00	

∗1 Thickness varies with frequency.

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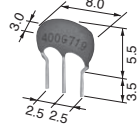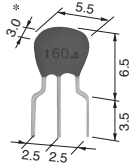
Characteristics and Types of CERALOCK® 1

1

■Dimensions and Standard Land Pattern of Chip CERALOCK® (CSTCC/CSTCR/CSTCE/CSTCW Series)

Part Number	Frequency (MHz)	Dimensions Standard Land Pattern (in mm)
CSTCC☐☐G*2	2.00–3.99	
CSTCR☐☐G*2	4.00–7.99	
CSTCE☐☐G*2	8.00–13.99	
CSTCE☐☐V*2	14.00–20.00	
CSTCW☐☐X*2	20.01–70.00	

*1 Thickness varies with frequency.
*2 Conformal coating or washing of the components is not acceptable because they are not hermetically sealed.

■Dimensions of Carrier Tape for Chip CERALOCK®

CSTCR Series

The cover film peel strength force 0.1 to 0.7N
The cover film peel speed 300mm/min.

Direction of Feed

(in mm)

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2 **Principles of CERALOCK®**

1. Equivalent Circuit Constants

Fig. 2-1 shows the symbol for a ceramic resonator. The impedance and phase characteristics measured between the terminals are shown in Fig. 2-2. This illustrates that the resonator becomes inductive in the frequency zone between the frequency Fr (resonant frequency), which provides the minimum impedance, and the frequency Fa (anti-resonant frequency), which provides the maximum impedance.

It becomes capacitive in other frequency zones. This means that the mechanical vibration of a two-terminal resonator can be replaced equivalently with a combination of series and parallel resonant circuits consisting of an inductor : L, a capacitor : C, and a resistor : R. In the vicinity of the specific frequency (Refer to Note 1 on page 8), the equivalent circuit can be expressed as shown in Fig. 2-3.

Fr and Fa frequencies are determined by the piezoelectric ceramic material and the physical parameters. The equivalent circuit constants can be determined from the following formulas. (Refer to Note 2 on page 8)

$$Fr = 1/2\pi \sqrt{L_1 C_1} \qquad\qquad\qquad (2\text{-}1)$$
$$Fa = 1/2\pi \sqrt{L_1 C_1 C_0/(C_1+C_0)} = Fr \sqrt{1+C_1/C_0} \qquad (2\text{-}2)$$
$$Qm = 1/2\pi Fr C_1 R_1 \qquad\qquad\qquad (2\text{-}3)$$

(Qm : Mechanical Q)

Considering the limited frequency range of $Fr \leqq F \leqq Fa$, the impedance is given as $Z = Re + j\omega Le$ $(Le \geqq 0)$ as shown in Fig. 2-4, and CERALOCK® should work as an inductance Le (H) having the loss Re (Ω).

Symbol ○───┤├───○

Impedance between Two Terminals Z=R+jx
(R : Real Component, X : Impedance Component)
Phase φ =tan⁻¹X/R
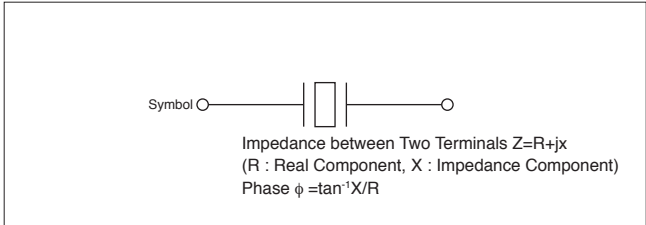
Fig. 2-1 Symbol for the Two-Terminal CERALOCK®
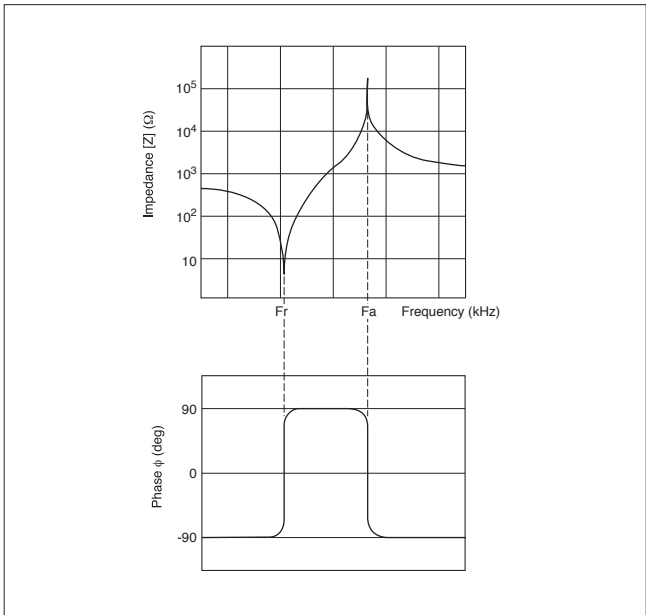
Fig. 2-2 Impedance and Phase Characteristics of CERALOCK®

R₁ : Equivalent Resistance
L₁ : Equivalent Inductance
C₁ : Equivalent Capacitance
C₀ : Parallel Equivalent Capacitance

Fig. 2-3 Electrical Equivalent Circuit of CERALOCK®

Re : Effective Resistance
Le : Effective Inductance
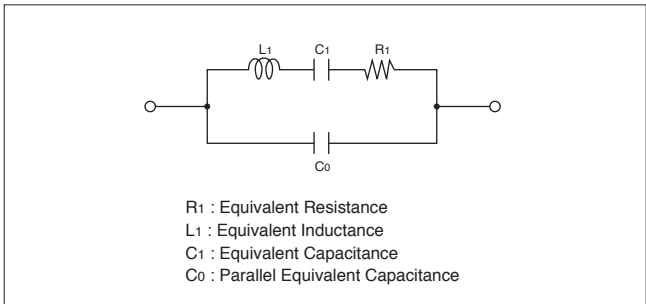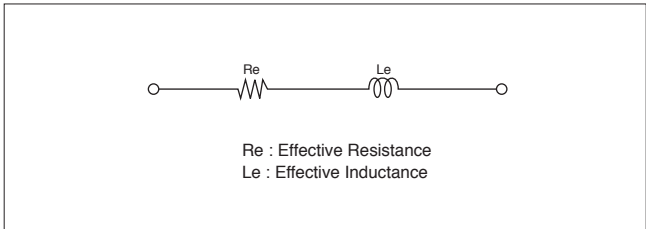
**Fig. 2-4 Equivalent Circuit of CERALOCK®
in the Frequency Band Fr≦F≦Fa**

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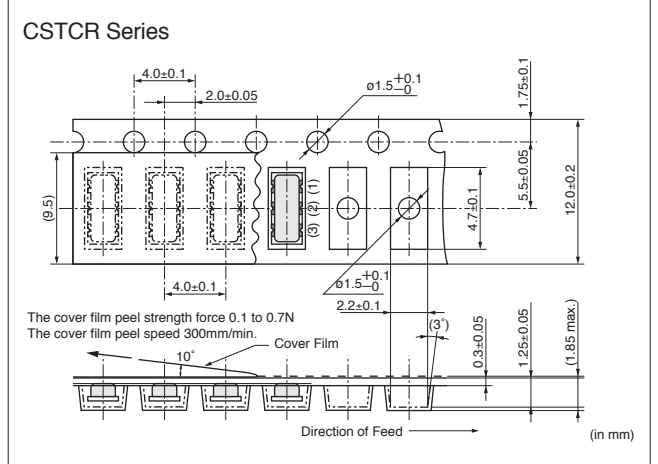
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Principles of CERALOCK® | 2

The table on this page shows a comparison of the equivalent constants between CERALOCK® and a quartz crystal oscillator.

In comparison, there is a large difference in capacitance and Qm, which results in the difference of oscillating conditions, when actually operated.

The table in the appendix shows the standard values of an equivalent circuit constant for each type of CERALOCK®. Furthermore, other higher harmonic modes exist, other than the desired oscillation mode. These other oscillation modes exist because the ceramic resonator uses mechanical resonance.

Fig. 2-5 shows those characteristics.
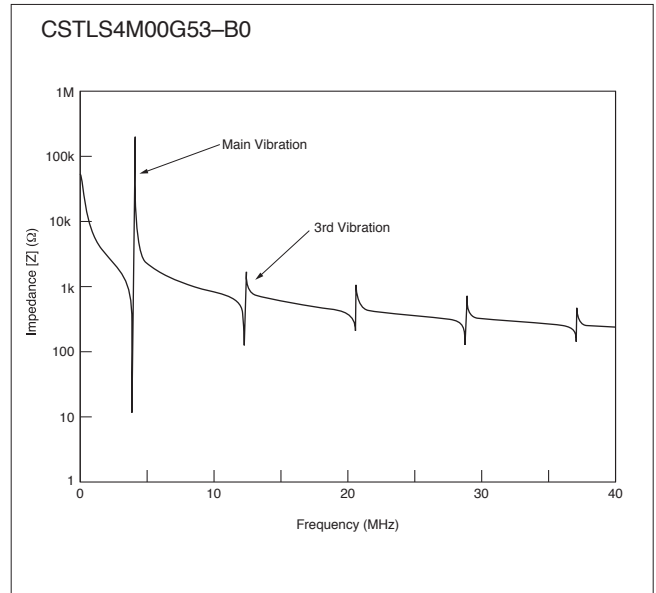
Fig. 2-5 Spurious Characteristics of CERALOCK®

■Comparison of Equivalent Circuits of CERALOCK® and Crystal Oscillator

Resonator	Oscillation Frequency	L_1 (μH)	C_1 (pF)	C_0 (pF)	R_1 (Ω)	Qm	dF (kHz)
CERALOCK®	2.00MHz	1.71×10^3	4.0	20.8	43.9	475	177.2
	4.00MHz	0.46×10^3	3.8	19.8	9.0	1220	350.9
	8.00MHz	0.13×10^3	3.5	19.9	8.0	775	641.6
Crystal	2.457MHz	7.2×10^5	0.005	2.39	37.0	298869	3
	4.00MHz	2.1×10^5	0.007	2.39	22.1	240986	6
	8.00MHz	1.8×10^5	0.002	4.48	154.7	59600	2

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2 Principles of CERALOCK®

2

Notes

(Note 1)

The relationship between the size of the resonator and the resonant frequency is described as follows. For example, the frequency doubles if the thickness doubles, when thickness vibration is used. The following relationship is obtained when the length of the resonators is ℓ, the resonance frequency is Fr, the speed of sound waves travelling through piezoelectric ceramics, and the wavelength is λ.

$$\begin{cases} \text{Fr} \cdot \ell = \text{Const.} \\ \text{(frequency constant, Fr·t for the thickness)} \\ \lambda = 2\,\ell \\ \text{C} = \text{Fr} \cdot \lambda = 2\text{Fr} \cdot \ell \end{cases}$$

As seen in the above formula, the frequency constant determines the size of the resonator.

$\ell = \lambda\,/\,2$

Amplitude Range of Standing Wave

(Min.Amplitude) (Max.Amplitude)
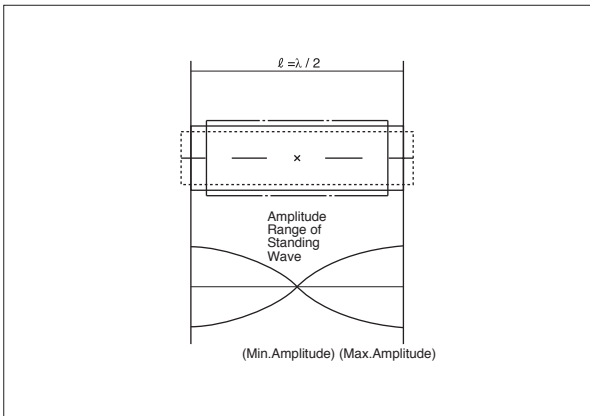
Fig. Ⅰ

(Note 2)

In Fig. 2-3, when resistance R_1 is omitted for simplification, the impedance $Z(\omega)$ between two terminals is expressed by the following formula.

$$Z(\omega) = \frac{\dfrac{1}{j\omega C_0}\left(j\omega L_1 + \dfrac{1}{j\omega C_1}\right)}{\dfrac{1}{j\omega C_0} + \left(j\omega L_1 + \dfrac{1}{j\omega C_1}\right)}$$

$$= \frac{j\left(\omega L_1 - \dfrac{1}{\omega C_1}\right)}{1 + \dfrac{C_0}{C_1} - \omega^2 C_0 L_1}$$

When $\omega = \dfrac{1}{\sqrt{L_1 C_1}} = \omega r$, $Z(\omega r) = 0$

When $\omega = \dfrac{1}{\sqrt{C_0 C_1 L_1/(C_0+C_1)}} = \omega a$, $Z(\omega a) = \infty$

Therefore from $\omega = 2\pi F$,

$$Fr = \omega r/2\pi = \frac{1}{2\pi\sqrt{L_1 C_1}}$$

$$Fa = \omega a/2\pi = \frac{1}{2\pi\sqrt{C_0 C_1 L_1/(C_0+C_1)}} = Fr\sqrt{1 + \frac{C_1}{C_0}}$$

L_1 C_1

C_0

Fig. Ⅱ

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Principles of CERALOCK® **2**

2. Basic Oscillation Circuits

Generally, basic oscillation circuits can be grouped into the following 3 categories.
① Use of positive feedback
② Use of negative resistance element
③ Use of delay in transfer time or phase
In the case of ceramic resonators, quartz crystal oscillators, and LC oscillators, positive feedback is the circuit of choice.
Among the positive feedback oscillation circuit using an LC, the tuning type anti-coupling oscillation circuit, Colpitts and Hartley circuits are typically used.
See Fig. 2-6.

In Fig. 2-6, a transistor, which is the most basic amplifier, is used.
The oscillation frequencies are approximately the same as the resonance frequency of the circuit consisting of L, C_{L1} and C_{L2} in the Colpitts circuit or consisting of L_1 and L_2 in the Hartley circuit. These frequencies can be represented by the following formulas. (Refer to Note 3 on page 11.)

(Colpitts Circuit)

$$fosc. \fallingdotseq \frac{1}{2\pi \sqrt{L \cdot \dfrac{C_{L1} \cdot C_{L2}}{C_{L1} + C_{L2}}}} \quad\text{————————}\quad (2\text{-}4)$$

(Hartley Circuit)

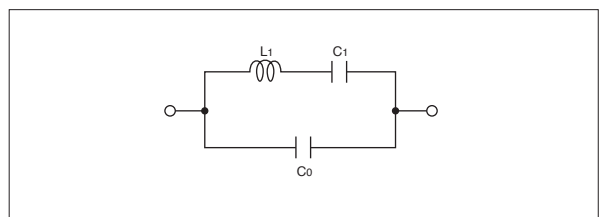$$fosc. \fallingdotseq \frac{1}{2\pi \sqrt{C (L_1+L_2)}} \quad\text{————————}\quad (2\text{-}5)$$

In an LC network, the inductor is replaced by a ceramic resonator, taking advantage of the fact that the resonator becomes inductive between resonant and anti-resonant frequencies.
This is most commonly used in the Colpitts circuit.
The operating principle of these oscillation circuits can be seen in Fig. 2-7. Oscillation occurs when the following conditions are satisfied.

$$\left\{ \begin{array}{l} \text{Loop Gain G} = \alpha \cdot \beta \geqq 1 \\ \text{Phase Amount} \hspace{3cm} (2\text{-}6) \\ \quad \theta = \theta_1 + \theta_2 = 360° \times n \ (n = 1, 2, \cdots) \end{array} \right.$$

In Colpitts circuit, an inverter of $\theta_1 = 180°$ is used, and it is inverted more than $\theta_2 = 180°$ with L and C in the feedback circuit. The operation with a ceramic resonator can be considered the same.

Colpitts Circuit   Hartley Circuit

Fig. 2-6 Basic Configuration of LC Oscillation Circuit

Amplifier
$\left(\begin{array}{l} \text{Mu Factor} : \alpha \\ \text{Phase Shift} : \theta_1 \end{array} \right)$

Feedback Circuit
$\left(\begin{array}{l} \text{Feedback Ratio} : \beta \\ \text{Phase Shift} : \theta_2 \end{array} \right)$

Oscillation Conditions
Loop Gain G= $\alpha \cdot \beta \geqq 1$
Phase Shift $\theta = \theta_1 + \theta_2 = 360° \times n$
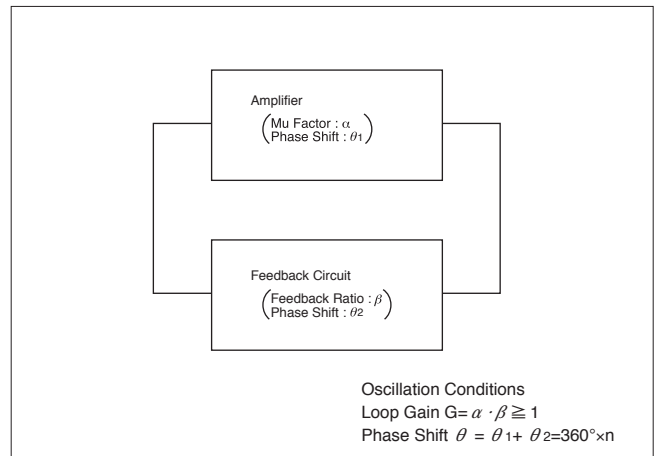
Fig. 2-7 Principle of Oscillation

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2 Principles of CERALOCK®

It is common and simple to utilize an inverter for the
Colpitts circuit with CERALOCK®.
Fig. 2-8 shows the basic oscillation circuit with inverter.
In an open loop circuit by cutting at point Ⓐ , it is
possible to measure loop gain G and phase shift θ.
Fig. 2-9 shows the actual measuring circuit, and an
example of the measuring result is shown in Fig. 2-10.
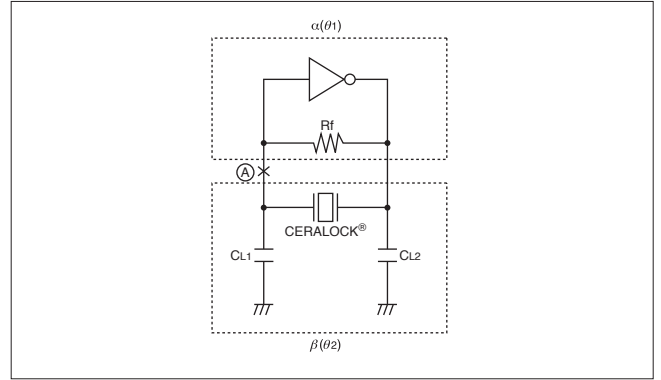
Fig. 2-8 Basic Oscillation Circuit with Inverters
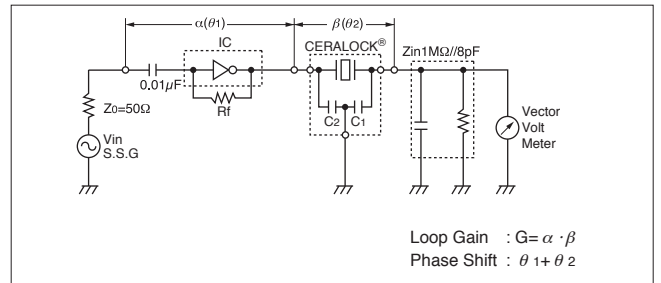
Fig. 2-9 Measuring Circuit Network of Loop Gain and Phase Shift

Loop Gain : $G = \alpha \cdot \beta$
Phase Shift : $\theta_1 + \theta_2$

CERALOCK®
CSTLS4M00G53–B0
V_{DD}=+5V
C_{L1}=C_{L2}=15pF
IC : TC4069UBP
(TOSHIBA)

CERALOCK®
CSTLS4M00G53–B0
V_{DD}=+2V
C_{L1}=C_{L2}=15pF
IC : TC4069UBP
(TOSHIBA)
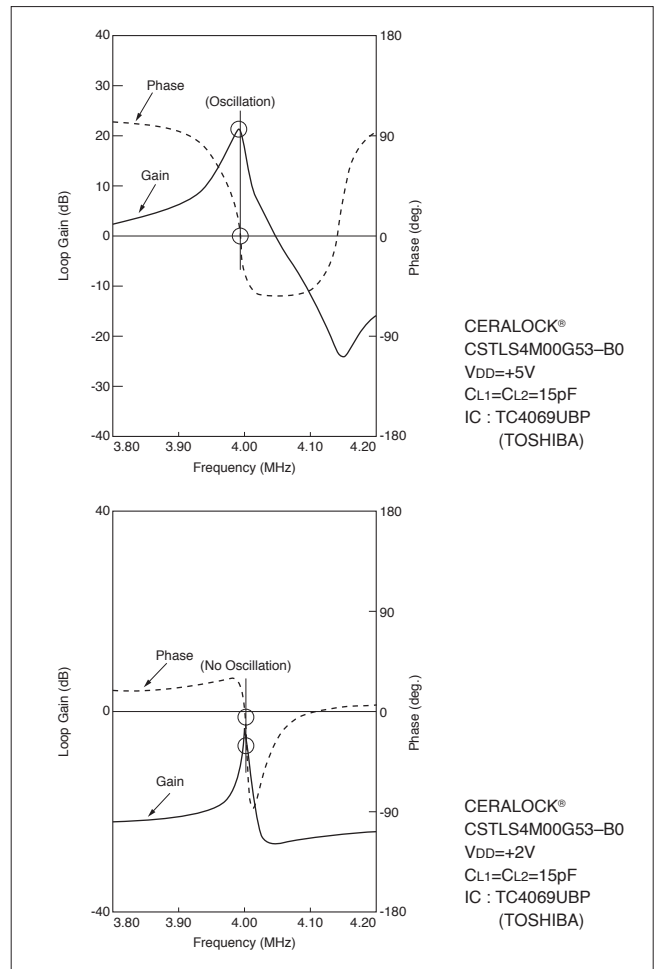
Fig. 2-10 Measured Results of Loop Gain and Phase Shift

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Principles of CERALOCK® 2

2

Notes

(Note 3)

Fig. III shows the equivalent circuit of an emitter grounding type transistor circuit. In the figure, Ri stands for input impedance, R$_0$ stands for output impedance and ß stands for current amplification rate.

When the oscillation circuit in Fig.2-6 is expressed by using the equivalent circuit in Fig. III, it becomes like Fig. IV. Z_1, Z_2 and Z are as shown in the table for each Hartley type and Colpitts type circuit. The following 3 formulas are obtained based on Fig. IV.

Fig. III

	Hartley Type	Colpitts Type
Z_1	$j\omega L_1$	$1 / j\omega C_{L1}$
Z_2	$j\omega L_2$	$1 / j\omega C_{L2}$
Z	$1 / j\omega C$	$j\omega L$

Fig. IV Hartley/Colpitts Type LC Oscillation Circuits

$$\begin{cases} \beta R_0 i_1 + (R_0+Z_2)\ i_2 - Z_2 i_3 = 0 & \cdots\cdots (1) \\ Z_1 i_1 + Z_2 i_2 - (Z_2+Z+Z_1)\ i_3 = 0 & \cdots\cdots (2) \\ (Z_1+Ri)\ i_1 - Z_1 i_3 = 0 & \cdots\cdots (3) \end{cases}$$

As $i_1 \neq 0$, $i_2 \neq 0$, $i_3 \neq 0$ are required for continuous oscillation, the following conditional formula can be performed by solving the formulas of (1), (2) and (3) on the current.
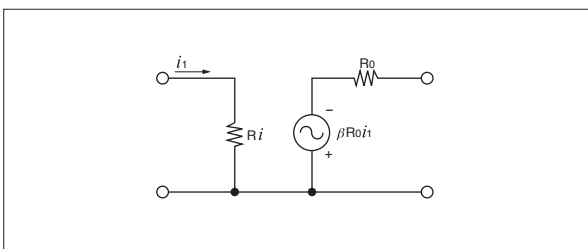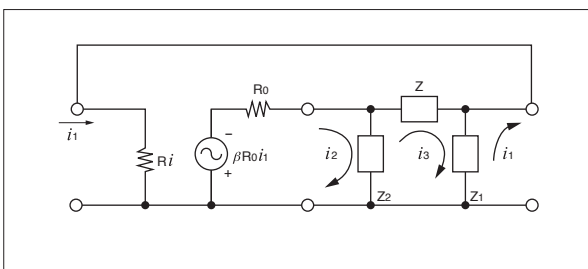
$$\begin{cases} \beta R_0 Z_1 Z_2 = (Z_1+Ri)Z_2^2 - \{Z_1(Z_2+Z)+ \\ \qquad (Z_2+Z+Z_1)Ri\}(Z_2+R_0) \qquad \cdots\cdots (4) \end{cases}$$

Then, as Z_1, Z_2 and Z are all imaginary numbers, the following conditional formula is obtained by dividing the formula (4) into the real number part and the imaginary number part.

$$\begin{cases} \text{(Imaginary number part)} \\ \quad Z_1 Z_2 Z + (Z_1+Z_2+Z)RiR_0 = 0 \qquad \cdots\cdots (5) \\ \text{(Real number part)} \\ \quad \beta R_0 Z_1 Z_2 + Z_1(Z+Z_2)R_0 + \\ \quad Z_2(Z+Z_1)Ri = 0 \qquad\qquad \cdots\cdots (6) \end{cases}$$

Formula (5) represents the phase condition and formula (6) represents the power condition. Oscillation frequency can be obtained by applying the elements shown in the aforementioned table to Z_1, Z_2 and Z solving it for angular frequency ω.

(Hartley Type)

$$\omega^2 osc = (2\pi\ fosc.)^2 = \cfrac{1}{(L_1 L_2)\ C\{1+ \cfrac{L_1 \cdot L_2}{(L_1+L_2)\ CRiR_0}\}}$$

$$\cdots\cdots (7)$$

(Colpitts Type)

$$\omega^2 osc = (2\pi\ fosc.)^2 = \cfrac{1}{L \cfrac{C_{L1} \cdot C_{L2}}{C_{L1}+C_{L2}}} \cdot \{1+ \cfrac{L}{(C_{L1}+C_{L2})\ RiR_0}\}$$

$$\cdots\cdots (8)$$

In either circuit, the term in brackets will be 1 as long as Ri and R$_0$ is large enough. Therefore oscillation frequency can be obtained by the following formula.

(Hartley Type) $fosc. \fallingdotseq \cfrac{1}{2\pi \sqrt{(L_1+L_2)\ C}}$ $\cdots\cdots (9)$

(Colpitts Type) $fosc. \fallingdotseq \cfrac{1}{2\pi \sqrt{L \cdot \cfrac{C_{L1} \cdot C_{L2}}{C_{L1}+C_{L2}}}}$ $\cdots (10)$

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3 Specifications of CERALOCK®

1. Electrical Specifications

The frequency stability of CERALOCK® is between that of crystal and LC or RC oscillators. Temperature stability is ±0.3 to ±0.5% against initial values within -20 to +80°C. The initial frequency precision is ±0.5% for standard products. The frequency of the standard CERALOCK® is adjusted by the standard measuring circuit, but the oscillation frequency may shift when used in the actual IC circuit. Usually, if the frequency precision needed for clock signal of a 1 chip microcomputer is approximately ±2 to 3% under working conditions, CERALOCK® standard type can be used in most cases. If exact oscillation frequency is required for a special purpose, Murata can manufacture the ceramic resonator for the desired frequency.
The following are the general electrical specifications of CERALOCK®. (As for the standard measuring circuit of oscillation frequency, please refer to the next chapter "Application to Typical Oscillation Circuits".)

Electrical Specifications of MHz Band Lead CERALOCK® (CSTLS Series)

Electrical specifications of CSTLS series are shown in the tables. Please note that oscillation frequency measuring circuit constants of the CSTLS □G56 series (with H-CMOS IC) depends on frequency.

■Resonant Impedance Specifications of CSTLS/ Series

Type	Frequency Range (MHz)	Resonant Impedance (Ω max.)
CSTLS□G	3.40 — 3.99	50
	4.00 — 7.99	30
	8.00 — 10.00	25
CSTLS□X	16.00 — 32.99	50
	33.00 — 50.00	40

MHz band three-terminal CERALOCK® (CSTLS Series) is built-in load capacitance.
Fig. 3-1 shows the electrical equivalent circuit.
The table shows the general specifications of the CSTLS series. Input and output terminals of the three-terminal CERALOCK® are shown in the table titled Dimensions of CERALOCK® CSTLS series in Chapter 1 on page 6. But connecting reverse, the oscillating characteristics are not affected except that the frequency has a slight lag.
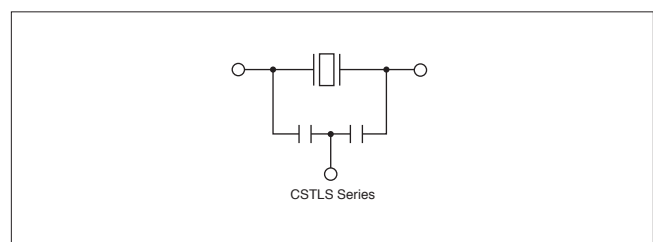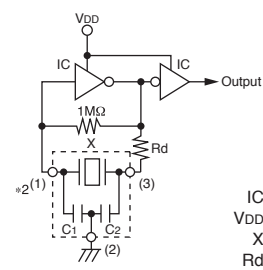
CSTLS Series

Fig. 3-1 Symbol for the Three-Terminal CERALOCK®

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Specifications of CERALOCK® 3

■**General Specifications CSTLS Series**

Part Number / Item	Frequency Range (MHz)	Initial Tolerance of Oscillation Frequency	Temperature Stability of Oscillation Frequency (-20 to +80°C)	Oscillating Frequency Aging	Standard Circuit for Oscillation Frequency
CSTLS☐G53/56	3.40—10.00	±0.5%	±0.2%[*1]	±0.2%	
CSTLS☐X	16.00—50.00	±0.5%	±0.2%	±0.2%	

∗1 This value varies for built-in Capacitance
∗2 If connected conversely, a slight frequency lag may occur.
∗3 G56/X series : TC74HCU04(TOSHIBA)
∗4 This resistance value applies to the CSTLS☐G56 series.

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3 Specifications of CERALOCK®

Electrical Specifications of MHz Band Chip CERALOCK® (CSACW Series) (CSTCC/CSTCR/CSTCE/CSTCW Series)

General specifications of chip CERALOCK® (CSACW series)and (CSTCC/CSTCR/CSTCE/CSTCW series) are shown in the tables respectively.

■Resonant Impedance of CSTCC/CSTCR/CSTCE/CST(A)CW Series

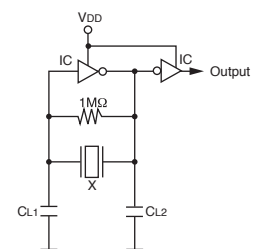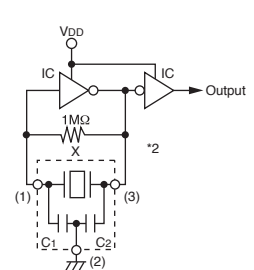| Type	Frequency Range (MHz)	Resonant Impedance (Ω max.)
CSTCC□G	2.00— 2.99	80
	3.00— 3.99	50
CSTCR□G	4.00— 5.99	60
	6.00— 7.99	50
CSTCE□G	8.00—10.00	40
	10.01—13.99	30
CSTCE□V	14.00—20.00	40
CSACW□X/CSTCW□X	20.01—24.99	80
	25.00—29.99	60
	30.00—60.00	50
	60.01—70.00	60

■General Specifications of CSACW Series

Item / Part Number	Frequency Range (MHz)	Initial Tolerance of Oscillation Frequency	Temperature Stability of Oscillation Frequency (-20 to +80°C)	Oscillating Frequency Aging	Standard Circuit for Oscillation Frequency
CSACW□X53	20.01—24.99	±0.5%	±0.2%	±0.1%	
CSACW□X51	25.00—70.00	±0.5%	±0.2%	±0.1%	IC : TC74HCU04*(TOSHIBA) V_DD : +5V X : Chip CERALOCK® C_L1, C_L2 : This value varies for frequency.

* X51 Series (60.01—70.00MHz); SN74AHCU04

■General Specifications of CSTCC/CSTCR/CSTCE/CSTCW Series

Item / Part Number	Frequency Range (MHz)	Initial Tolerance of Oscillation Frequency	Temperature Stability of Oscillation Frequency (-20 to +80°C)	Oscillating Frequency Aging	Standard Circuit for Oscillation Frequency
CSTCC□G	2.00— 3.99	±0.5%	±0.3%*3	±0.3%	
CSTCR□G	4.00— 7.99	±0.5%	±0.2%	±0.1%	
CSTCE□G	8.00—13.99	±0.5%	±0.2%	±0.1%	
CSTCE□V	14.00—20.00	±0.5%	±0.3%	±0.3%	IC : TC4069UBP*1(TOSHIBA) V_DD : +5V X : Chip CERALOCK®
CSTCW□X	20.01—70.00	±0.5%	±0.2%	±0.1%	

*1 V, X Series; TC74HCU04(TOSHIBA), X Series (50.00—70.00MHz); SN74AHCU04(TI)
*2 If connected in the wrong direction, the above specification may not be guaranteed.
*3 This value varies for built-in Capacitance and Frequency.

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Specifications of CERALOCK® **3**

2. Mechanical and Environmental Specifications of CERALOCK®

The tables show the standard test conditions of mechanical strength and environmental specifications of CERALOCK®.
Fig. 3-2 shows the changes of oscillation frequency in each test, the table on the next page shows the criteria after the tests, and Fig. 3-3 shows the reflow soldering profile.

■Test Conditions for Standard Reliability of CERALOCK®

Item	Conditions
1. Shock Resistance	Measure after dropping from a height of ⬚a⬚ cm to ⬚b⬚ floor surface 3 times.
2. Soldering Heat Resistance	Lead terminals are immersed up to 2.0 mm from the resonator's body in solder bath of ⬚c⬚, and then the resonator shall be measured after being placed in natural condition for 1 hour.[*1] Reflow profile show in Fig. 3-3 of heat stress is applied to the resonator, then the resonator shall be measured after being placed in natural condition for 1 hour.[*2]
3. Vibration Resistance	Measure after applying vibration of 10 to 55Hz amplitude of 2 mm to each of 3 directions, X, Y, Z.
4. Humidity Resistance	Keep in a chamber with a temperature of ⬚d⬚ and humidity of 90 to 95% for ⬚e⬚ hours. Leave for 1 hour before measurement.
5. Storage at High Temperature	Keep in a chamber at 85±2°C for ⬚e⬚ hours. Leave for 1 hour before measurement.
6. Storage at Low Temperature	Keep in a chamber at ⬚f⬚ °C for ⬚e⬚ hours. Leave for 1 hour before measurement.
7. Temperature Cycling	Keep in a chamber at -55°C for 30 minutes. After leaving at room temperature for 15 minutes, keep in a chamber at +85°C for 30 minutes, and then room temperature for 15 minutes. After 10 cycles of the above, measure at room temperature.
8. Terminal Strength	Apply 1 kg of static load vertically to each terminal and measure.

∗1 Applies to CERALOCK® Lead Type
∗2 Applies to MHz Band Chip CERALOCK®

1. CSTLS Series

Type	fosc.	a	b	c	d	e	f
G	3.40—10.00MHz	100	concrete	350±10°C	60±2°C	1000	−55±2°C
X	16.00—50.00MHz	100	concrete	350±10°C	60±2°C	1000	−55±2°C

2. CSACW Series

Type	fosc.	a	b	c	d	e	f
X	20.01—50.00MHz	100	wooden plate	—	60±2°C	1000	−55±2°C

3. CSTCC/CSTCR/CSTCE/CSTCW Series

Type	fosc.	a	b	c	d	e	f
G	2.00—13.99MHz	100	wooden plate	—	60±2°C	1000	−55±2°C
V	14.00—20.00MHz	100	wooden plate	—	60±2°C	1000	−55±2°C
X	20.01—70.00MHz	100	wooden plate	—	60±2°C	1000	−55±2°C

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3 Specifications of CERALOCK®

1. Shock Resistance

2. Solder Heat Resistance

3. Vibration Resistance

4. Humidity Resistance

5. Storage at High Temperature

6. Storage at Low Temperature

7. Temperature Cycling

8. Terminal Strength

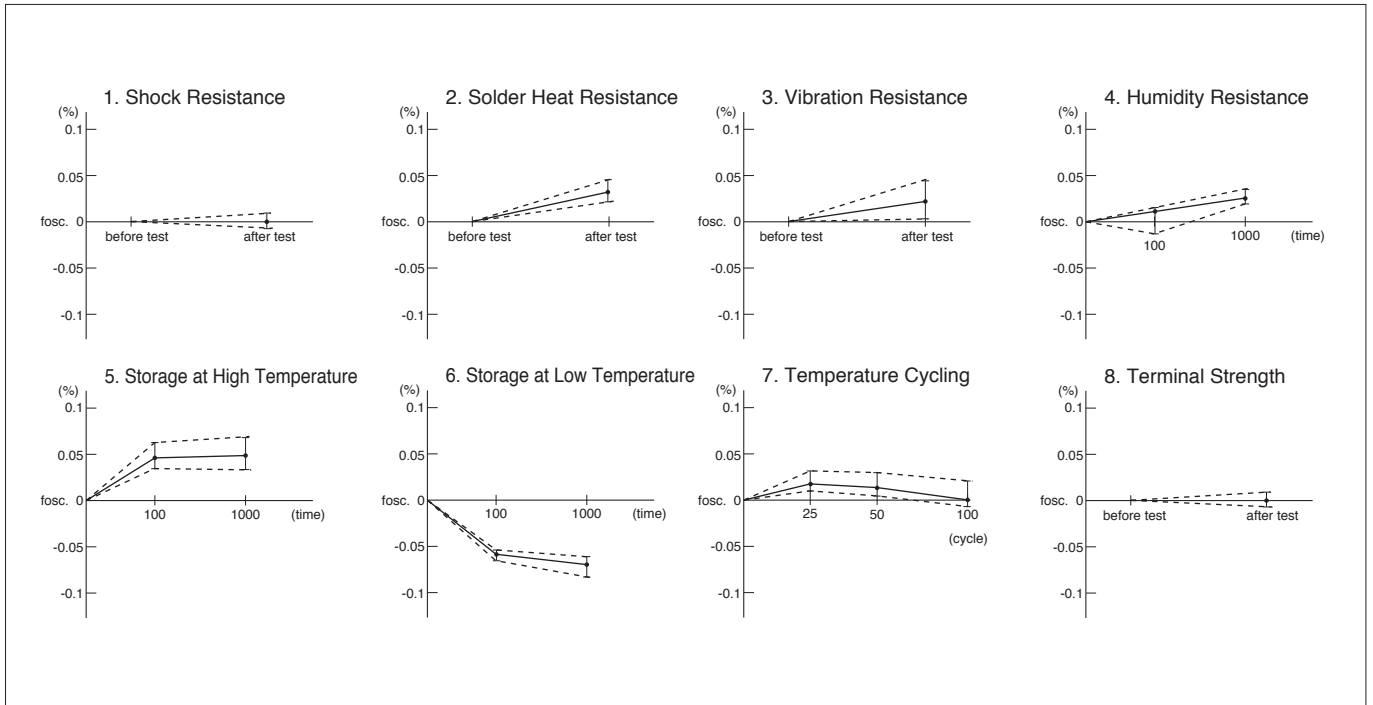Fig. 3-2 General Changes of Oscillation Frequency in Each Reliability Test (CSTLS4M00G53–B0)

■Deviation after Reliability Test

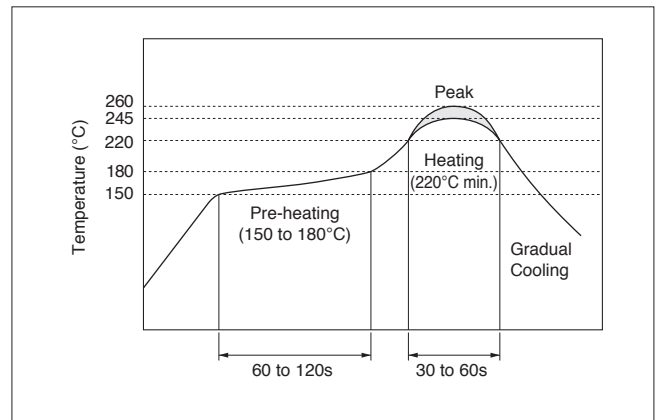|	Item	
Type	Oscillation Frequency	Other
Every Series	within±0.2%* (from initial value)	Meets the individual specification of each product.

∗ CSTCC Series : within±0.3%

Fig. 3-3 Reflow Soldering Profile for MHz Band Chip CERALOCK®

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4 Applications of Typical Oscillation Circuits

As described in Chapter 2, the most common oscillation circuit with CERALOCK® is to replace L of a Colpitts circuit with CERALOCK®. The design of the circuit varies with the application and the IC being used, etc. Although the basic configuration of the circuit is the same as that of a quartz crystal, the difference in mechanical Q results in the difference of the circuit constant.

This chapter briefly describes the characteristics of the oscillation circuit and gives some typical examples.

1. Cautions for Designing Oscillation Circuits

It is becoming more common to configure the oscillation circuit with a digital IC, and the simplest way is to use an inverter gate.

Fig. 4-1 shows the configuration of a basic oscillation circuit with a C-MOS inverter.

INV. 1 works as an inverter amplifier of the oscillation circuit. INV. 2 acts to shape the waveform and also acts as a buffer for the connection of a frequency counter.

The feedback resistance Rf provides negative feedback around the inverter in order to put it in the linear region, so the oscillation will start, when power is applied.

If the value of Rf is too large, and if the insulation resistance of the input inverter is accidentally decreased, oscillation will stop due to the loss of loop gain. Also, if Rf is too great, noise from other circuits can be introduced into the oscillation circuit.

Obviously, if Rf is too small, loop gain will be low. An Rf of 1MΩ is generally used with a ceramic resonator.

Damping resistor Rd provides loose coupling between the inverter and the feedback circuit and decreases the loading on the inverter, thus saving energy.

In addition, the damping resistor stabilizes the phase of the feedback circuit and provides a means of reducing the gain in the high frequency area, thus preventing the possibility of spurious oscillation.

Load capacitance C_{L1} and C_{L2} provide the phase lag of 180°.

The proper selected value depends on the application, the IC used, and the frequency.
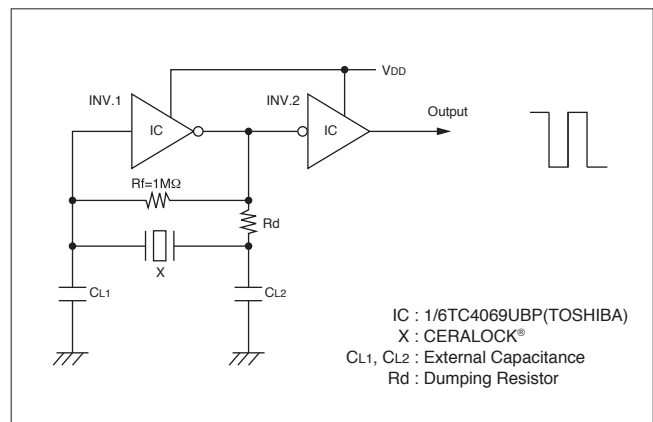

Fig. 4-1 Basic Oscillation Circuit with C-MOS Inverter

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4 Application to Typical Oscillation Circuits

Oscillation frequency fosc. in this circuit is expressed approximately by the following equation.

$$\text{fosc.}=\text{Fr}\sqrt{1+\frac{C_1}{C_0+C_L}} \hspace{2cm} (4\text{-}1)$$

Where, Fr=Resonance frequency of CERALOCK®

C_1 : Equivalent series capacitance of CERALOCK®

C_0 : Equivalent parallel capacitance of CERALOCK®

$$C_L=\frac{C_{L1}\ C_{L2}}{C_{L1}+C_{L2}}$$

This clearly shows that the oscillation frequency is influenced by the loading capacitance. Further caution should be paid in defining its value when a tight tolerance of oscillation frequency is required.

2. Application to Various Oscillation Circuits

Application to C-MOS Inverter

For the C-MOS inverting amplifier, the one-stage 4069 C-MOS group is best suited.
The C-MOS 4049 type is not used, because the three-stage buffer type has excessive gain, which causes RC oscillation and ringing.
Murata employs the TOSHIBA TC4069UBP as a C-MOS standard circuit. This circuit is shown in Fig. 4-2. The oscillation frequency of the standard CERALOCK® (C-MOS specifications) is adjusted by the circuit in Fig. 4-2.
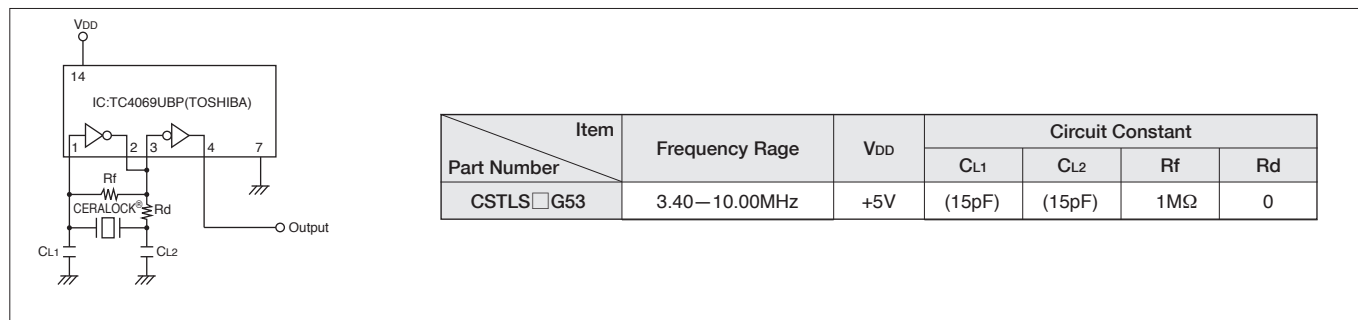
Item / Part Number	Frequency Rage	V_{DD}	Circuit Constant			
			C_{L1}	C_{L2}	Rf	Rd
CSTLS□G53	3.40—10.00MHz	+5V	(15pF)	(15pF)	1MΩ	0

Fig. 4-2 C-MOS Standard Circuit

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Application to Typical Oscillation Circuits 4

Application to H-CMOS Inverter

Recently, high-speed C-MOS (H-CMOS) have been used more frequently for oscillation circuits allowing high speed and energy saving control for the microprocessor. There are two types of H-CMOS inverters: the un-buffered 74HCU series and the 74HC series with buffers.

The 74HCU system is optimum for the CERALOCK® oscillation circuit.

Fig. 4-3 shows our standard H-CMOS circuit.

Since H-CMOS has high gain, especially in the high frequency area, greater loading capacitor (C_L) and damping resistor (Rd) should be employed to stabilize oscillation performance. As a standard circuit, we recommend Toshiba's TC74CU04, but any 74HCU04 inverter from other manufacturers may be used.

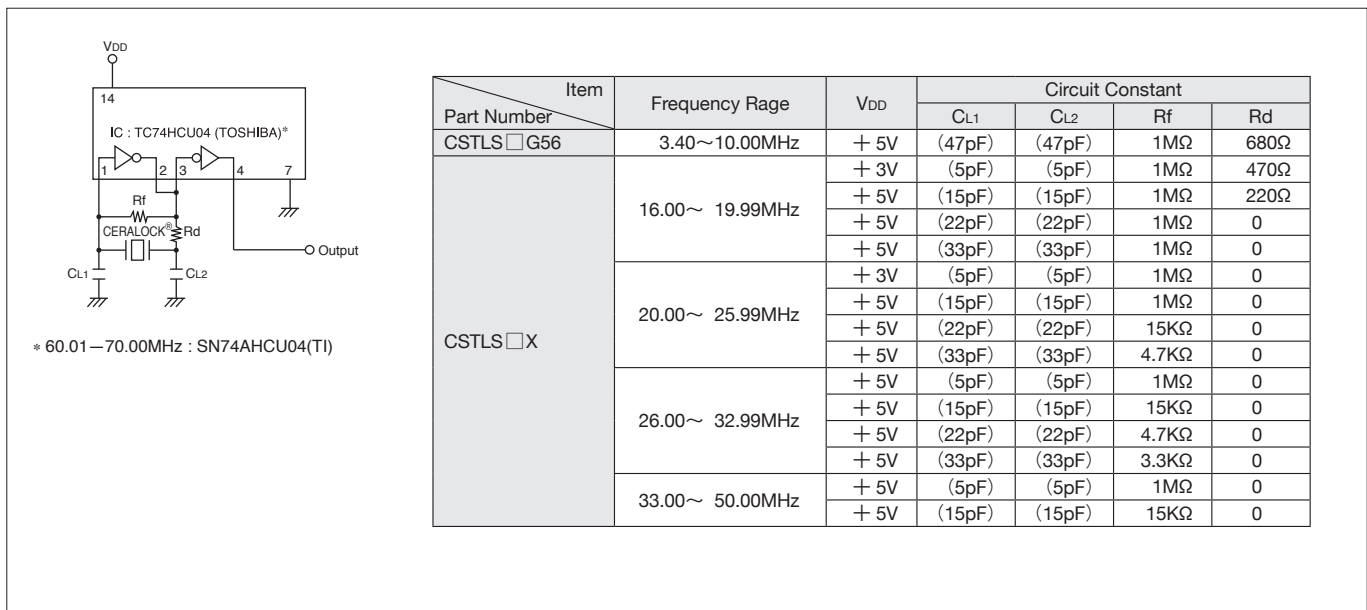The oscillation frequency for H-CMOS specifications is adjusted by the circuit in Fig. 4-3.

IC : TC74HCU04 (TOSHIBA)*

∗ 60.01―70.00MHz : SN74AHCU04(TI)

Part Number	Item Frequency Rage	V_{DD}	Circuit Constant			
			C_{L1}	C_{L2}	Rf	Rd
CSTLS□G56	3.40～10.00MHz	＋5V	(47pF)	(47pF)	1MΩ	680Ω
CSTLS□X	16.00～ 19.99MHz	＋3V	(5pF)	(5pF)	1MΩ	470Ω
		＋5V	(15pF)	(15pF)	1MΩ	220Ω
		＋5V	(22pF)	(22pF)	1MΩ	0
		＋5V	(33pF)	(33pF)	1MΩ	0
	20.00～ 25.99MHz	＋3V	(5pF)	(5pF)	1MΩ	0
		＋5V	(15pF)	(15pF)	1MΩ	0
		＋5V	(22pF)	(22pF)	15KΩ	0
		＋5V	(33pF)	(33pF)	4.7KΩ	0
	26.00～ 32.99MHz	＋5V	(5pF)	(5pF)	1MΩ	0
		＋5V	(15pF)	(15pF)	15KΩ	0
		＋5V	(22pF)	(22pF)	4.7KΩ	0
		＋5V	(33pF)	(33pF)	3.3KΩ	0
	33.00～ 50.00MHz	＋5V	(5pF)	(5pF)	1MΩ	0
		＋5V	(15pF)	(15pF)	15KΩ	0

Fig. 4-3 H-CMOS Standard Circuit

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5 Characteristics of CERALOCK® Oscillation Circuits

This chapter describes the general characteristics of the basic oscillation of Fig. 4-1 (page17). Contact Murata for detailed characteristics of oscillation with specific kinds of ICs and LSIs.

1. Stability of Oscillation Frequency

Fig. 5-1 shows examples of actual measurements for stability of the oscillation frequency.
The stability versus temperature change is ±0.1 to 0.5% within a range of -20 to +80°C, although it varies slightly depending on the ceramic material.
Influence of load capacitance (C_{L1}, C_{L2}) on the oscillation frequency is relatively high, as seen in formula (4-1) (Page18). It varies approximately ±0.05% for a capacitance deviation of ±10%. The stability versus supply voltage is normally within ±0.05% in the working voltage range, although it varies with the characteristics of the IC.
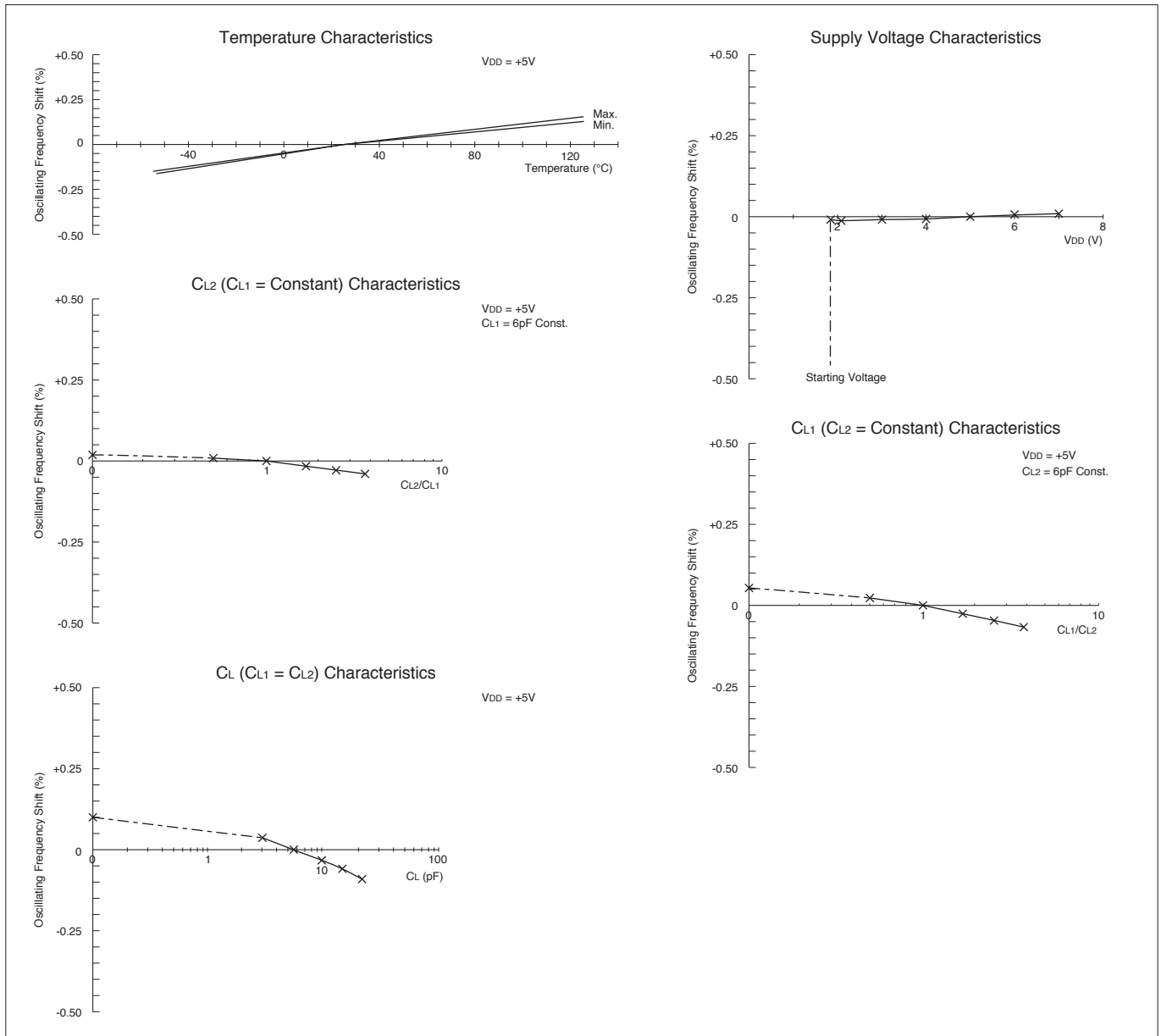

Fig. 5-1 Examples of Actual Measurement for the Stability of Oscillation Frequency (IC: TC74HCU04 (TOSHIBA), CERALOCK®: CSACW33M8X51–B0)

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Characteristics of CERALOCK® Oscillation Circuits   5

2. Characteristics of the Oscillation Level

Fig. 5-2 shows examples of actual measurements of the oscillation level versus temperature, supply voltage and load capacitance (C_{L1}, C_{L2}). The oscillating amplitude is required to be stable over a wide temperature range, and temperature characteristics should be as flat as possible. The graph titled Supply Voltage Characteristics in Fig. 5-2 shows that the amplitude varies linearly with supply voltage, unless the IC has an internal power supply voltage regulator.
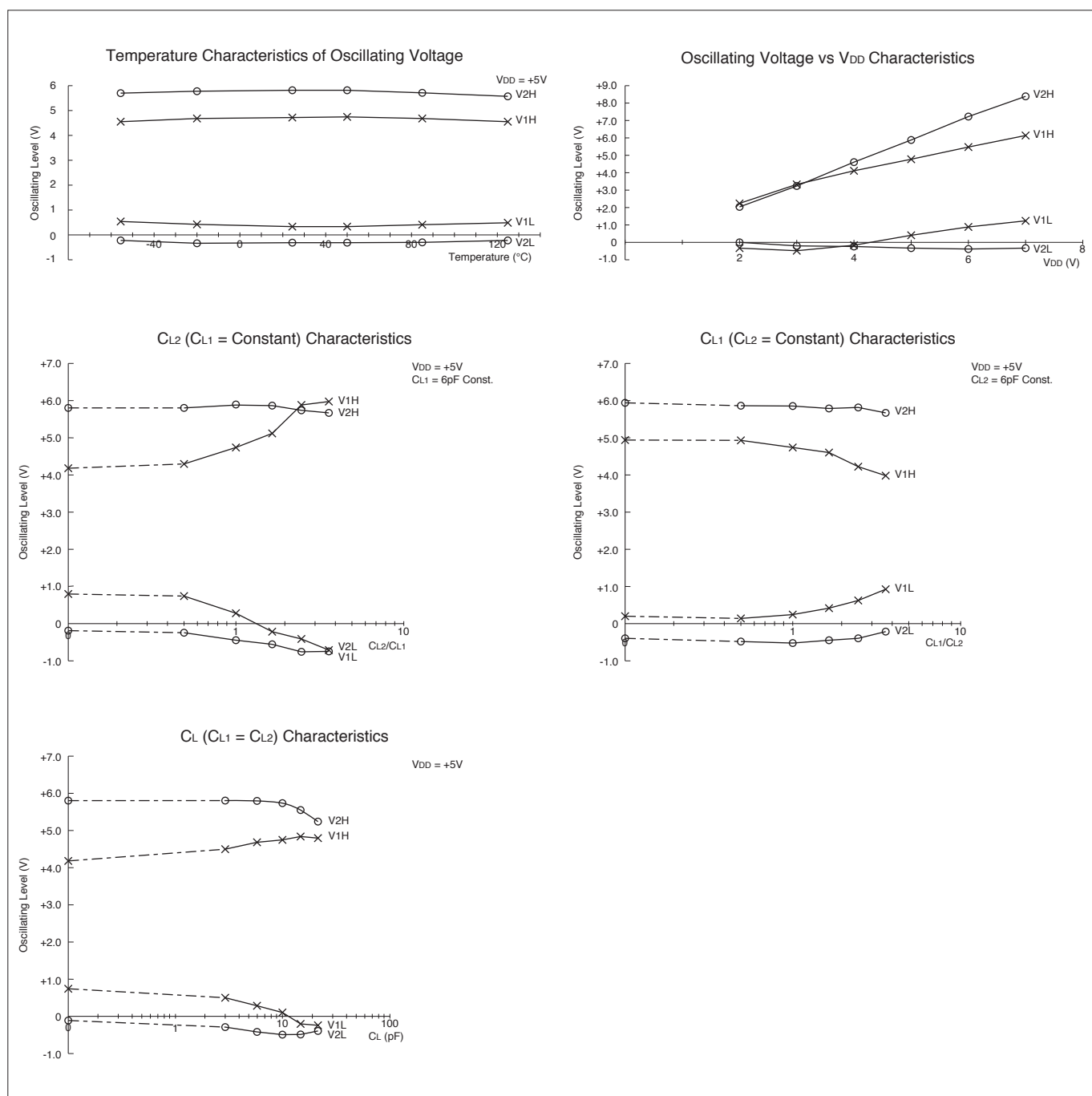

Fig. 5-2 Examples of Actual Measurement of Oscillating Amplitude (IC: TC74HCU04(TOSHIBA), CERALOCK®: CSACW33M8X51–B0)

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5 | Characteristics of CERALOCK® Oscillation Circuit

3. Characteristics of Oscillation Rise Time

Oscillation rise time means the time when oscillation develops from a transient area to a steady state condition, at the time the power of the IC is activated. With a CERALOCK®, this is defined as the time to reach 90% of the oscillation level under steady state conditions as shown in Fig. 5-3.

Rise time is primarily a function of the oscillation circuit design. Generally, smaller loading capacitance, higher frequency of ceramic resonator, and lower mechanical Q of ceramic resonator cause a faster rise time. The effect of load capacitance becomes more apparent as the capacitance of the resonator decreases.

Fig. 5-4 shows how the rise time increases as the load capacitance of the resonator increases. Also, Fig. 5-4 shows how the rise time varies with supply voltage.

It is noteworthy that the rise time of the ceramic resistor is one or two decades faster than a quartz crystal.

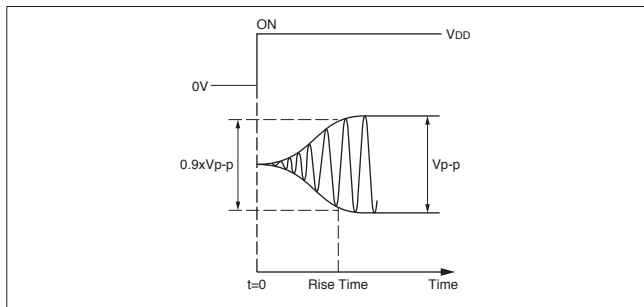Fig. 5-5 shows comparison of rise time between the two.
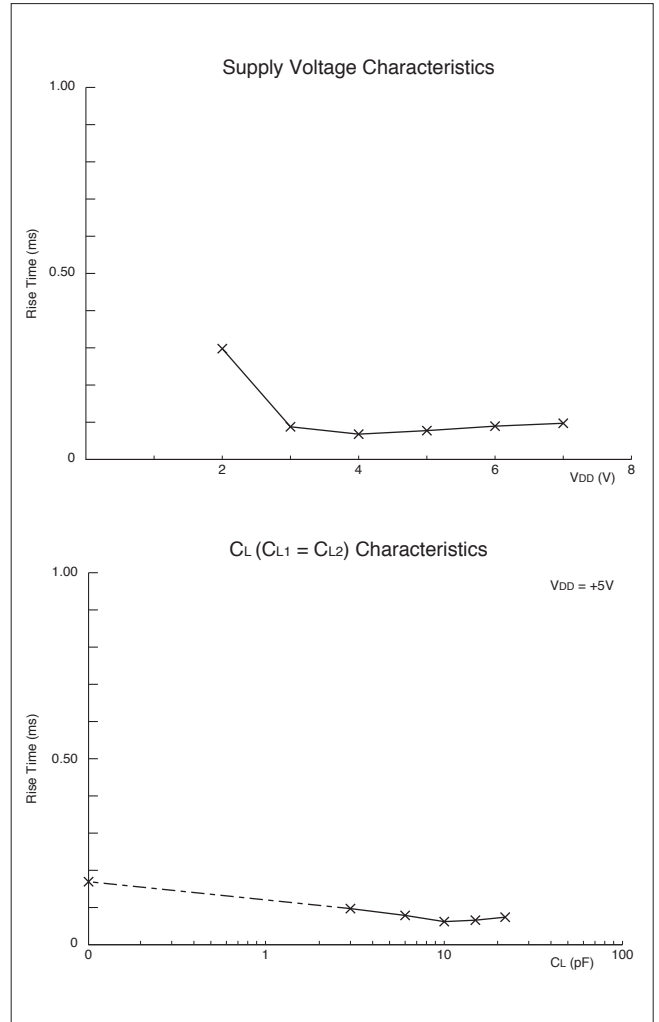

Fig. 5-3 Definition of Rise Time

Fig. 5-4 Examples of Characteristics of Oscillation Rise Time
(IC: TC74HCU04 (TOSHIBA),
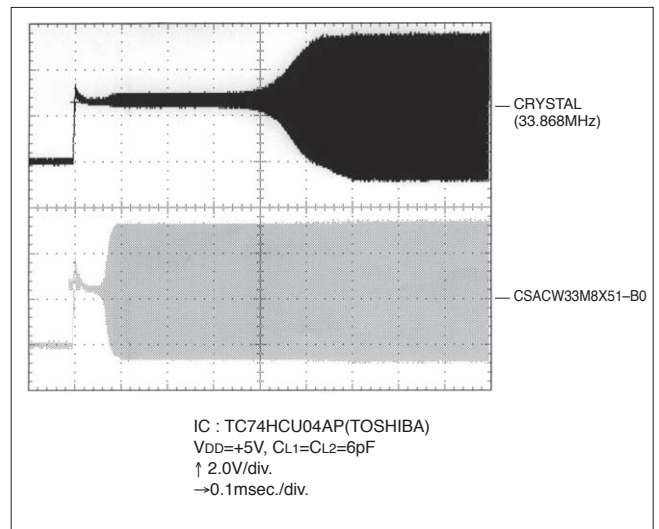CERALOCK®: CSACW33M8X51–B0)

IC : TC74HCU04AP(TOSHIBA)
V_{DD}=+5V, C_{L1}=C_{L2}=6pF
↑ 2.0V/div.
→0.1msec./div.

Fig. 5-5 Comparison of the Rise Time of
a Ceramic Resonator vs. a Quartz Crystal

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Characteristics of CERALOCK® Oscillation Circuits 5

4. Starting Voltage

Starting voltage refer to the minimum supply voltage at which an oscillation circuit can operate. Starting voltage is affected by all the circuit elements, but it is determined mostly by the characteristics of the IC. Fig. 5-6 shows an example of an actual measurement for the starting voltage characteristics against the loading capacitance.
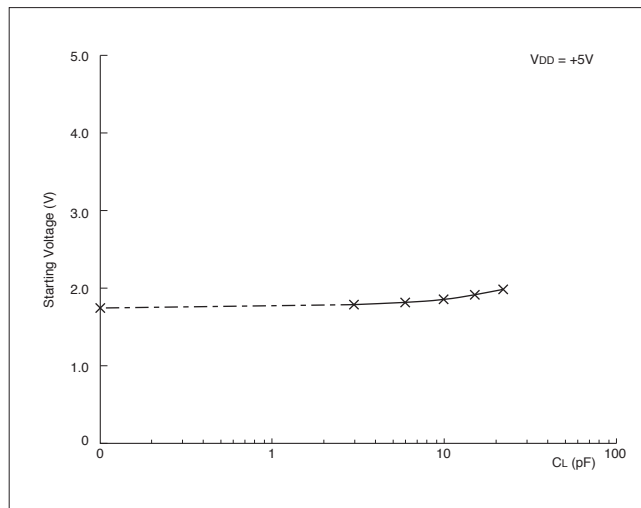

Fig. 5-6 Starting Voltage Characteristics against C_L ($C_{L1}=C_{L2}$)
(IC: TC74HCU04 (TOSHIBA), CERALOCK®:
CSACW33M8X51–B0)

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6 Application Circuits to Various ICs/LSIs

CERALOCK®, by making good use of the above-mentioned features, is used in a wide range of applications to various kinds of ICs.
The following are a few examples of actual applications.

1. Application to Microcomputers

CERALOCK® is optimum for a stable oscillation element for various kinds of microcomputers : 4-bit, 8-bit and 16-bit.
With the general frequency tolerance required for the reference clock of microcomputers at ±2 to ±3%, standard CERALOCK® meets this requirement. Please consult with MURATA or LSI manufacturers about the circuit constants, because these constants vary with frequency and the LSI circuit being used.
Fig. 6-1 to 6-5 show applications to various kinds of 4-bit microcomputers, Fig. 6-6 to 6-8 show application to 8-bit microcomputers, and Fig. 6-9 to 6-10 show application to 16bit and 32bit microcomputers.

The recomended circuit condition of many ICs has been uploaded to Murata Web site. Please access to the below URL.
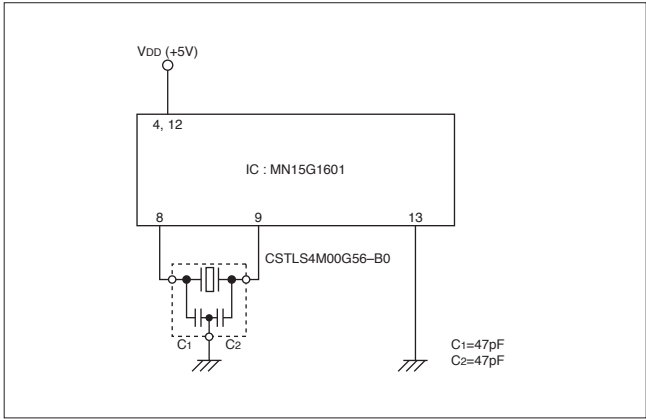http://search.murata.co.jp/Ceramy/ICsearchAction.do?sLang=en

Fig. 6-1 Application to MN15G1601 (Panasonic)
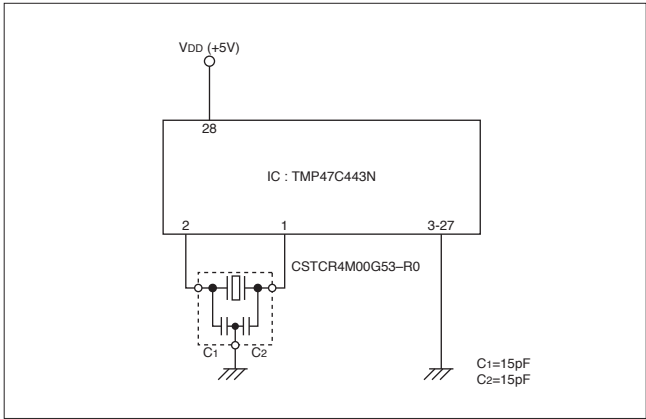
Fig. 6-2 Application to TMP47C443N (TOSHIBA)
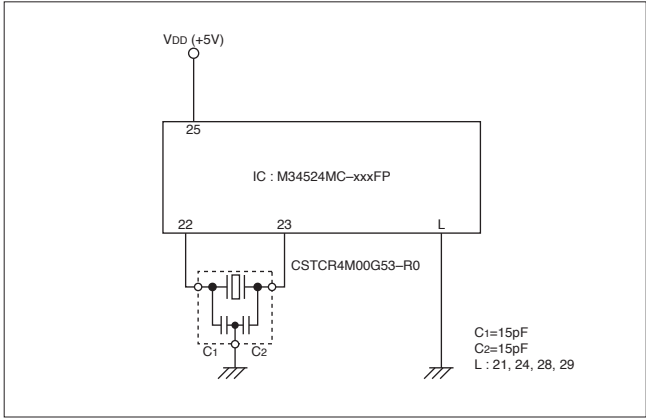
Fig. 6-3 Application to M34524MC-xxxFP
(Renesas Electronics)

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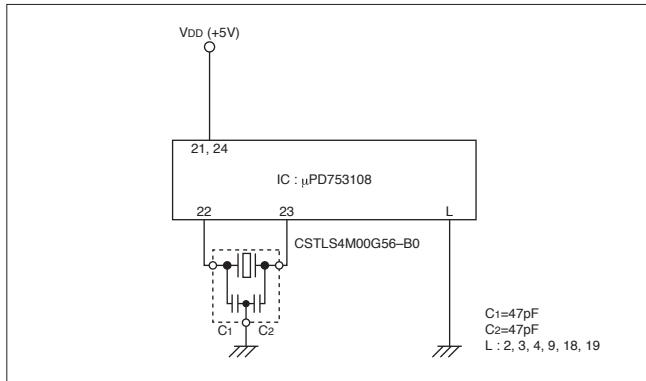
Application Circuits to Various ICs/LSIs **6**

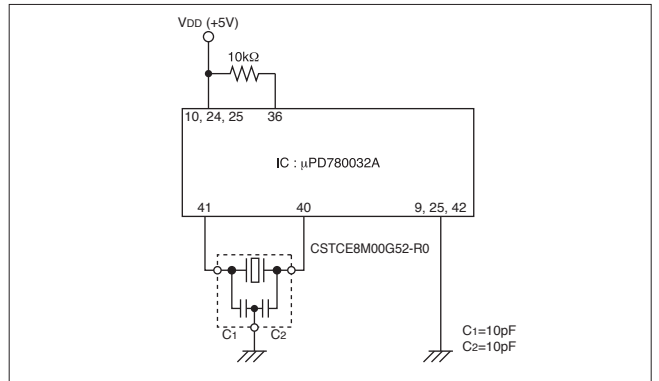Fig. 6-4 Application to μPD753108 (Renesas Electronics)

Fig. 6-7 Application to μPD780032A (Renesas Electronics)
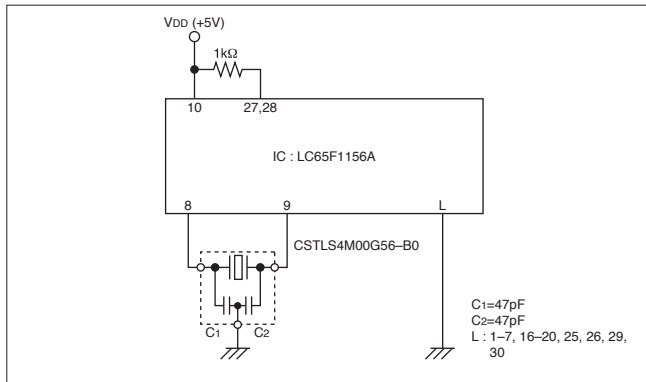

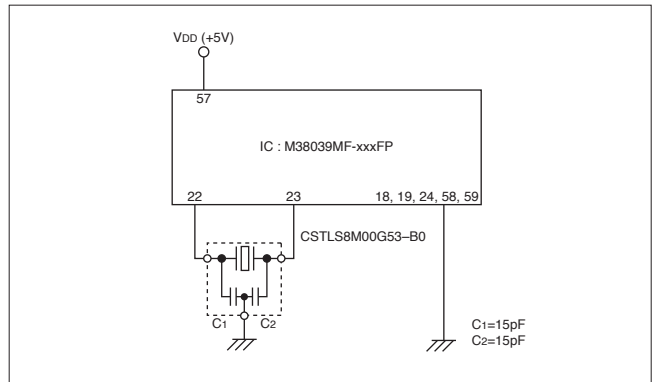Fig. 6-5 Application to LC65F1156A (SANYO)

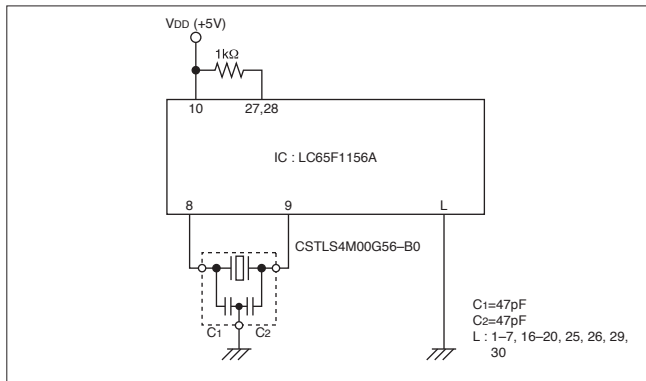Fig. 6-8 Application to M38039MF-xxxFP
(Renesas Electronics)

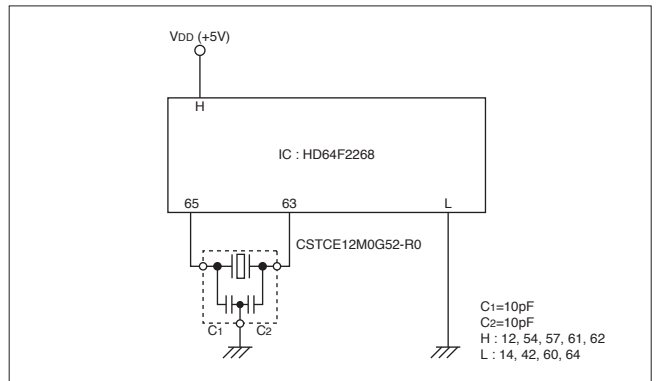Fig. 6-6 Application to TMP87C809BN (TOSHIBA)

Fig. 6-9 Application to HD64F2268
(Renesas Electronics)

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6 Application Circuits to Various ICs/LSIs

Fig. 6-10 Application to M30221M4-xxxFP
(Renesas Electronics)

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Application Circuits to Various ICs/LSIs **6**

2. Application to Remote Control ICs

Remote controll have become an increasingly more popular feature in TVs, stereos, VCRs, and air conditioners.
Fig. 6-11 shows an example of CERALOCK® in remote control transmission ICs. Oscillation frequency is normally 3.2M to 4MHz, with 3.64MHz being the most popular. This 3.64MHz is divided by a carrier signal generator, so that a carrier of approximately 38kHz is generated.
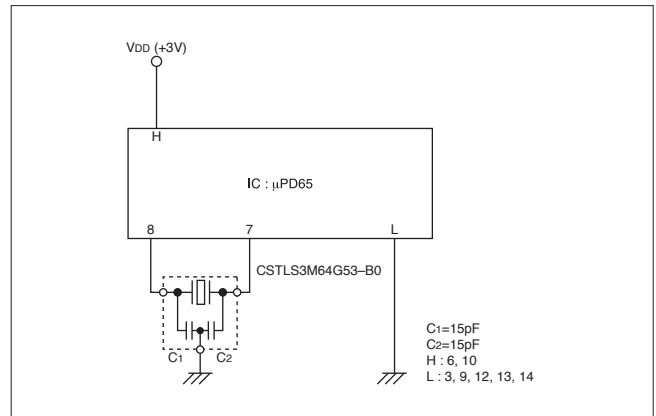

Fig. 6-11 Application to μPD65 (Renesas Electronics)

3. Application to ICs for Office Equipment

With the applications of ICs in office machines, many CERALOCK®s are used for motor drivers/controllers/digital signal processor (D.S.P.) in CD's ICs. Fig. 6-12 shows application example. It is believed that this type of application will increase in the future.
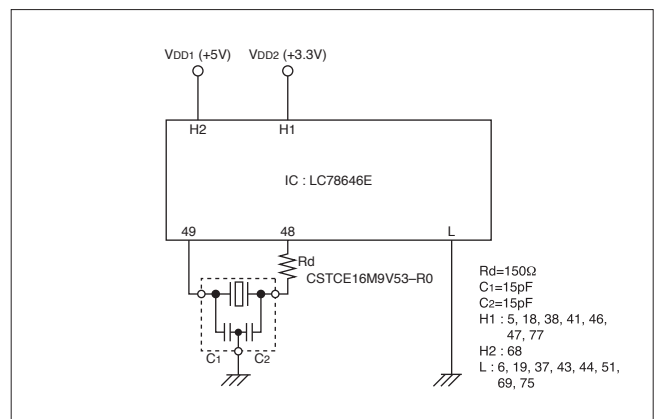

Fig. 6-12 Application to LC78646E (SANYO)
(CD Digital Signal Processor)

4. Other Kinds of Applications to Various ICs

Other than the above-mentioned uses, CERALOCK® is widely used with ICs for voice synthesis.
Fig. 6-13 shows an example of voice synthesis.
We can provide CERALOCK® application data for many ICs that are not mentioned in this manual. Please consult us for details.
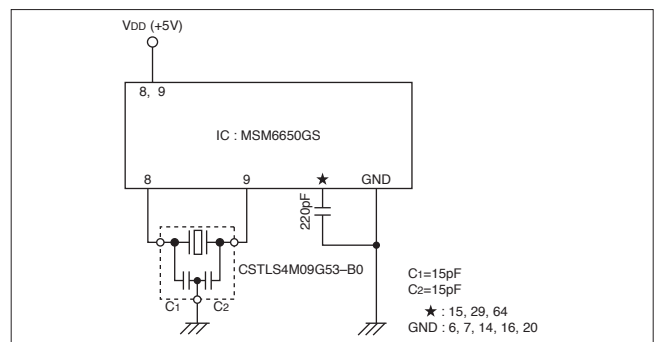

Fig. 6-13 Application to ICs for Voice Synthesis MSM6650GS (OKI)

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7 Notice

■Notice (Soldering and Mounting)

Please contact us regarding ultrasonic cleaning conditions to avoid possible damage.

■Notice (Storage and Operating Conditions)

Please do not apply excess mechanical stress to the component and lead terminals at soldering.

■Notice (Rating)

The component may be damaged if excess mechanical stress is applied.

■Notice (Handling)

· Unstable oscillation or oscillation stoppage might occur when CERALOCK® is used in an improper way in conjunction with ICs. We are happy to evaluate the application circuit to help you avoid this.

· Oscillation frequency of our standard CERALOCK® is adjusted with our standard measuring circuit. There could be slight shift in frequency if other types of IC are used. When you require exact oscillation frequency in your application, please contact us.

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8 Appendix Equivalent Circuit Constants of CERALOCK®

(The equivalent circuit constants are not the guaranteed value but the standard value.)

（MHz band lead CERALOCK®）

Part Number / Equivalent Constant	Fr (kHz)	Fa (kHz)	ΔF (kHz)	R_1 (Ω)	L_1 (mH)	C_1 (pF)	C_0 (pF)	Qm
CSTLS4M00G53-B0	3784.4	4135.3	350.9	9.0	0.4611	3.8377	19.7730	1220
CSTLS6M00G53-B0	5710.9	6199.5	488.6	7.5	0.2381	3.2635	18.2899	1135
CSTLS8M00G53-B0	7604.7	8246.3	641.6	8.0	0.1251	3.5030	19.9175	775
CSTLS10M0G53-B0	9690.1	10399.1	709.0	7.0	0.0984	2.7448	18.0899	947
CSTLS16M0X55-B0	15972.9	16075.0	102.1	24.6	0.6572	0.1511	11.7835	2681
CSTLS20M0X53-B0	19959.2	20070.8	111.6	19.0	0.4858	0.1309	11.6716	3203
CSTLS24M0X53-B0	23955.8	24095.9	140.2	16.6	0.4205	0.1050	8.9440	3805
CSTLS27M0X51-B0	27024.3	27172.8	148.5	15.9	0.3638	0.0953	8.6486	3877
CSTLS32M0X51-B0	31918.4	32092.6	174.2	13.4	0.2481	0.1002	9.1542	3716
CSTLS33M8X51-B0	33777.8	33969.7	191.9	25.6	0.2561	0.0867	7.6093	2120
CSTLS36M0X51-B0	36033.6	36241.1	207.6	13.4	0.2260	0.0863	7.4700	3821
CSTLS40M0X51-B0	39997.7	40240.1	242.7	15.8	0.2301	0.0688	5.6544	3651
CSTLS50M0X51-B0	49946.3	50193.1	246.8	27.6	0.1856	0.0547	5.5234	2107

（MHz band Chip CERALOCK®）

Part Number / Equivalent Constant	Fr (kHz)	Fa (kHz)	ΔF (kHz)	R_1 (Ω)	L_1 (mH)	C_1 (pF)	C_0 (pF)	Qm
CSTCC2M00G53-R0	1894.2	2092.8	198.6	16.1	1.8473	3.8235	17.3264	1375
CSTCR4M00G53-R0	3856.0	4098.6	242.6	16.0	0.8445	2.0176	15.5455	1304
CSTCR6M00G53-R0	5789.4	6152.4	363.0	11.9	0.3899	1.9396	14.9946	1207
CSTCE8M00G52-R0	7726.6	8177.4	450.8	7.5	0.2621	1.6201	13.4902	1715
CSTCE10M0G52-R0	9602.0	10172.0	570.0	7.2	0.1674	1.6477	13.4755	1401
CSTCE12M0G52-R0	11597.4	12285.0	687.6	5.8	0.1175	1.6023	13.1239	1483
CSTCE16M0V53-R0	15634.2	16574.4	940.2	10.4	0.1084	0.9563	7.7184	1039
CSTCE20M0V53-R0	19576.0	20761.0	1185.0	11.0	0.0791	0.8366	6.7052	932
CSTCW24M0X51-R0	23938.7	24090.8	152.1	24.1	0.4716	0.0938	7.3546	2953
CSTCW33M8X51-R0	33799.3	34003.7	204.4	24.8	0.3249	0.0683	5.6326	2789
CSTCW48M0X51-R0	47949.9	48227.0	277.1	23.0	0.1978	0.0557	4.8049	2609

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   ⑦ Traffic signal equipment   ⑧ Disaster prevention / crime prevention equipment
   ⑨ Data-processing equipment   ⑩ Application of similar complexity and/or reliability requirements to the applications listed above

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